

7-1 頻率響應分析方法

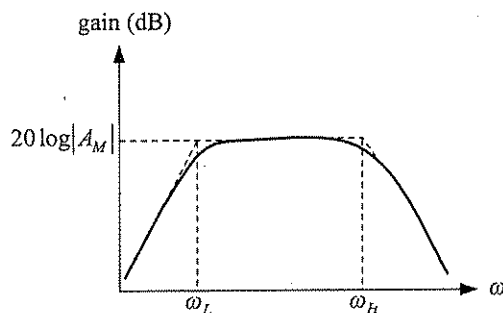


轉移函數與 Bode Plot

範例 1



The magnitude response of a capacitively coupled amplifier is shown in Fig.. Explain the reason for the magnitude decreasing at low and high frequencies.

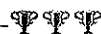


[91 中山通訊所 (甲乙)]

【解析】

放大器外部耦合或旁路電容 (大電容) 造成低頻衰減, 電晶體元件內部小電容造成高頻衰減。

範例 2



Find the midband gain in dB and bandwidth in Hz for the amplifier described

by $A(S) = \frac{2.5 \times 10^7 (S + 2 \times 10^5)}{(S + 10^5)(S + 5 \times 10^5)}$. What type of amplifier is this?

[97 成大微電子 / 電機 (甲乙丁戊) / 電通 (丙丁) 所]

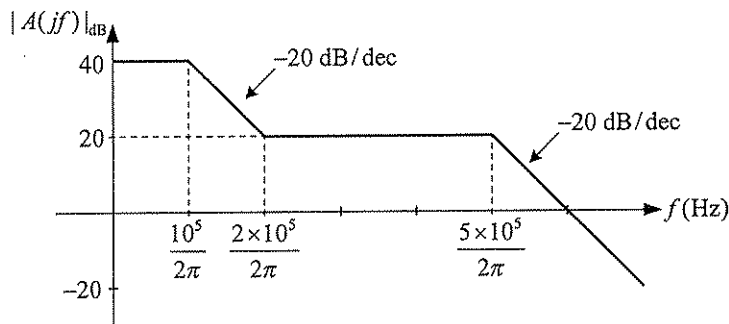
【解析】

將轉移函數標準化可得

$$A(jf) = \frac{100(1 + \frac{j2\pi f}{2 \times 10^5})}{(1 + \frac{j2\pi f}{10^5})(1 + \frac{j2\pi f}{5 \times 10^5})}$$

7-4 電子學經典題型解析 (II)

則大小波德圖為



由圖可知，中頻增益為40 dB，頻寬約為

$$f_H = \frac{1}{2\pi \sqrt{\left(\frac{1}{10^5}\right)^2 + \left(\frac{1}{5 \times 10^5}\right)^2 - 2\left(\frac{1}{2 \times 10^5}\right)^2}} = 21.658 \text{ (KHz)}$$

此為低通放大器。

範例 3

For the transfer function:

$$A_v(S) = \frac{8 \times 10^{13} S(S+20)}{(S+30)(S+400)(S+10000)(S+100000)}$$

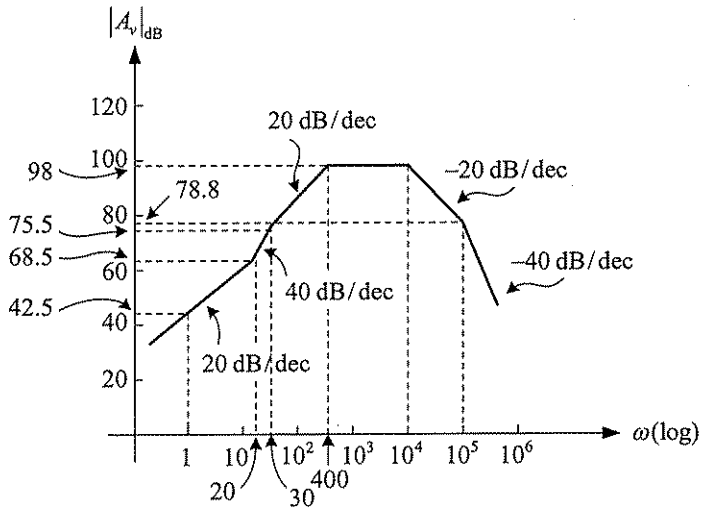
- (1) please write down the poles and zeros.
- (2) please state which pole is the dominant low frequency pole and which frequency is the dominant high frequency pole.
- (3) please sketch the Bode magnitude plot. [95成大奈米科技暨微系統所]

【解析】

- (1) zeros : $z_1 = 0$ (rad/sec) , $z_2 = -20$ (rad/sec) , $z_3 = z_4 = \infty$ (rad/sec)
 poles : $p_1 = -30$ (rad/sec) , $p_2 = -400$ (rad/sec) , $p_3 = -10^4$ (rad/sec) ,
 $p_4 = -10^5$ (rad/sec)

(2)&(3)將轉移函數表為波德圖標準式可得

$$A_v(S) = \frac{\frac{400}{3} S \left(1 + \frac{S}{20}\right)}{\left(1 + \frac{S}{30}\right) \left(1 + \frac{S}{400}\right) \left(1 + \frac{S}{10^4}\right) \left(1 + \frac{S}{10^5}\right)}$$



由主極點近似法可知

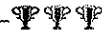
$$f_L = \frac{400}{2\pi} = 63.66 \text{ (Hz)}$$

$$f_H = \frac{10^4}{2\pi} = 1.59 \text{ (KHz)}$$



極點頻率、零點頻率與 3 dB 頻率求法

範例 4



The high-frequency response of an amplifier is characterized by 2 zeros at $S = \infty$ and 2 poles frequency at ω_{p1} and ω_{p2} . For $\omega_{p2} = k\omega_{p1}$, find the value of k that results in the exact value of ω_H being $0.9\omega_{p1}$.

[97海洋固態電子所]

【解析】

由於

$$\frac{1}{\omega_H} = \frac{1}{\omega_{p1}} + \frac{1}{\omega_{p2}} \Rightarrow \frac{1}{0.9\omega_{p1}} = \frac{1}{\omega_{p1}} + \frac{1}{k\omega_{p1}}$$

可得 $k = 9$

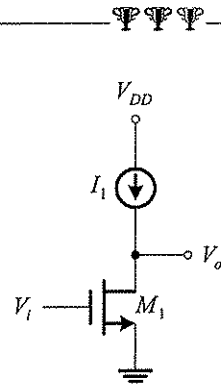
範例 5

右圖電路是一個 Common-Source Amplifier with Active Load。電晶體 M_1 及電流源 I_1 具有相同的 Output Resistance, r_o 。令 V_i 訊號源的輸出阻抗為 0, 則此放大器的 transfer function 可寫成：

$$A_v(S) = \frac{V_o(S)}{V_i(S)} = A_M \frac{1 - S/\omega_z}{1 + S/\omega_p}$$

求：

- (1) A_M 。
- (2) ω_z 。
- (3) ω_p 之公式，以 g_m 、 r_o 、 C_{gs} 、 C_{gd} 表示。



[95 交大電子所]

【解析】

由輸出端之節點方程式： $(SC_{gd} + 2/r_o)V_o(S) = (SC_{gd} - g_m)V_i(S)$ ，可得

$$\frac{V_o(S)}{V_i(S)} = A_M \frac{1 - S/\omega_z}{1 + S/\omega_p}$$

$$(1) \quad A_M = -g_m(r_o // r_o) = -\frac{1}{2}g_m r_o \text{ (V/V)}$$

(2) 由輸出端列節點方程式

$$V_o \left(\frac{1}{r_o} + \frac{1}{r_o} + SC_{gd} \right) = (SC_{gd} - g_m)V_{gs}$$

令 V_o 為零且 $S = z$ ，可得

$$z = \frac{g_m}{C_{gd}}$$

因此

$$\omega_z = |z| = \frac{g_m}{C_{gd}} \text{ (rad/sec)}$$

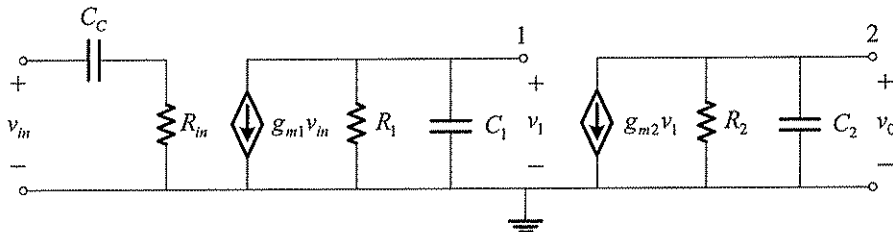
$$(3) \quad \omega_p = \frac{1}{\tau} = \frac{2}{C_{gd}r_o}$$

範例 6



The small-signal equivalent circuit of a two-stage amplifier is shown in the Fig., where $R_{in} = 1\text{ M}\Omega$, $C_C = 1\text{ }\mu\text{F}$, $g_{m1} = 100\text{ }\mu\text{A/V}^2$, $g_{m2} = 125\text{ }\mu\text{A/V}^2$, $R_1 = 200\text{ K}\Omega$, $R_2 = 250\text{ K}\Omega$, $C_1 = 0.5\text{ pF}$ and $C_2 = 2.5\text{ pF}$.

- (1) Determine the midband gain (in dB).
- (2) If a compensation capacitor (2 pF) is added between nodes 1 and 2, estimate the dominant pole frequency (in Hz).
- (3) Roughly sketch the magnitude response of the amplifier gain (v_o/v_{in}) after compensation, and indicate the -3 dB frequencies (in Hz) in your plot.
- (4) Find the zero results from the compensation capacitor. Is the zero located in the right-half or the left-half plane?



[95大同電機所]

【解析】

- (1) 中頻分析： C_C short, C_1 、 C_2 open

$$A_M = \frac{V_o}{V_{in}} = \frac{V_1}{V_{in}} \frac{V_o}{V_1} = (-g_{m1}R_1)(-g_{m2}R_2) = 625$$

因此

$$A_M(\text{dB}) = 20 \log |A_M| = 56(\text{dB})$$

- (2) 由於

$$\frac{V_o}{V_1} = -g_{m2}R_2 = -31.25 \triangleq K$$

所以